

## ECE 717

### Homework 2: State space solutions

due: Wednesday October 2, 2019

1. **Discretizing continuous equation.** In practice it is difficult to work with continuous-time signals because of storage and computational considerations. In this problem, we will discretize continuous state-space equations in a way that preserves the continuous-time behavior.

- a) Consider the continuous-time state-space system:  $\dot{x}(t) = Ax(t)$  with initial condition  $x(0) = x_0$ . We would like to find an exact discretization of this system. Specifically, we will sample  $x(t)$  at times  $t = 0, h, 2h, \dots$ . To this effect, define the discrete-time signal

$$x_d[k] = x(kh) \quad \text{for } k = 0, 1, \dots$$

Find a matrix  $A_d$  such that  $x_d[k+1] = A_d x_d[k]$ . The matrix  $A_d$  should depend on  $A$  and  $h$ .

- b) Let's add an input signal. Suppose we have the discrete input  $u_d[k]$ , and define the continuous input signal  $u(t)$  being constant between sampling times. In other words:

$$u(t) = u_d[k] \quad \text{for } kh \leq t < (k+1)h$$

Now consider the continuous-time system:  $\dot{x}(t) = Ax(t) + Bu(t)$  using the piecewise-constant input defined above. Find matrices  $A_d$  and  $B_d$  such that  $x_d[k+1] = A_d x_d[k] + B_d u_d[k]$ . Here,  $x_d[k]$  is defined the same as in part (a). You may express  $B_d$  as an integral.

- c) Use MATLAB to simulate  $x(t)$  for part (a) for the case where

$$A = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -5 & -9 & -5 \end{bmatrix}$$

Plot all three components of  $x(t)$  for  $0 \leq t \leq 5$ , with  $x_0 = [1 \ 1 \ 1]^T$ . Then, simulate the discretized system using  $h = 0.2$ . Verify that the two simulations agree (the second one should be a sampled version of the first one). Note: the MATLAB command for computing matrix exponentials is `expm`.

#### SOLUTION:

- a) The solution to the autonomous system is  $x(t) = e^{At}x_0$ . Substituting  $t = kh$  and using the definition for  $x_d[k]$ , we obtain

$$x_d[k+1] = e^{A(k+1)h}x_0 = e^{Ah}e^{Akh}x_0 = e^{Ah}x_d[k]$$

So the discretized version has  $A_d = e^{Ah}$ .

- b) The solution to the autonomous system at time  $t_2$  as a function of the initial time  $t_1$  is given by  $x(t_2) = e^{A(t_2-t_1)}x(t_1) + \int_{t_1}^{t_2} e^{A(t_2-\tau)}Bu(\tau) d\tau$ . Substituting  $t_1 = kh$  and  $t_2 = (k+1)h$  and using the definition for  $x_d[k]$ , we obtain

$$\begin{aligned} x_d[k+1] &= e^{Ah}x_d[k] + \int_{kh}^{kh+h} e^{A(kh+h-\tau)}Bu(\tau) d\tau \\ &= e^{Ah}x_d[k] + \left( \int_{kh}^{kh+h} e^{A(kh+h-\tau)}B d\tau \right) u_d[k] \\ &= e^{Ah}x_d[k] + \left( \int_0^h e^{A\tau}B d\tau \right) u_d[k] \end{aligned}$$

So the discretized version has  $A_d = e^{Ah}$  and  $B_d = \int_0^h e^{A\tau} B d\tau$ .

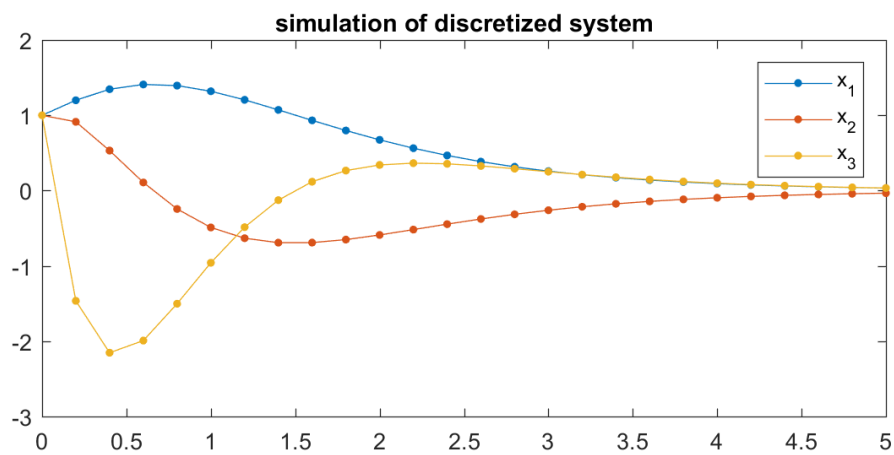
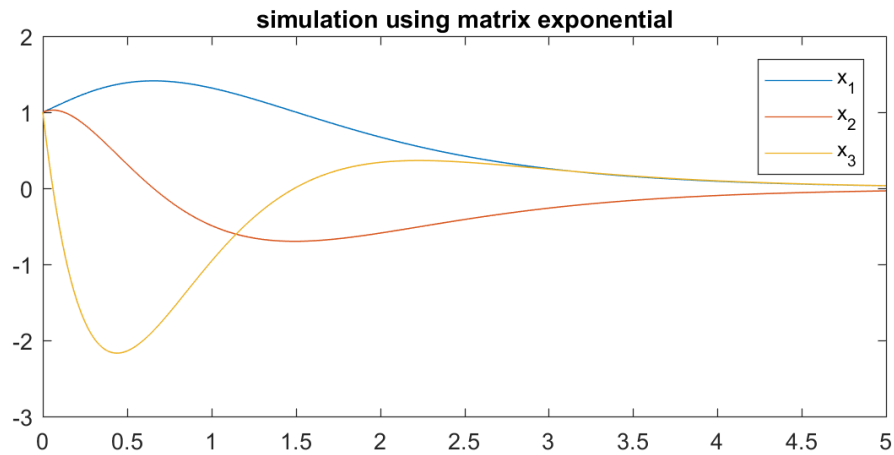
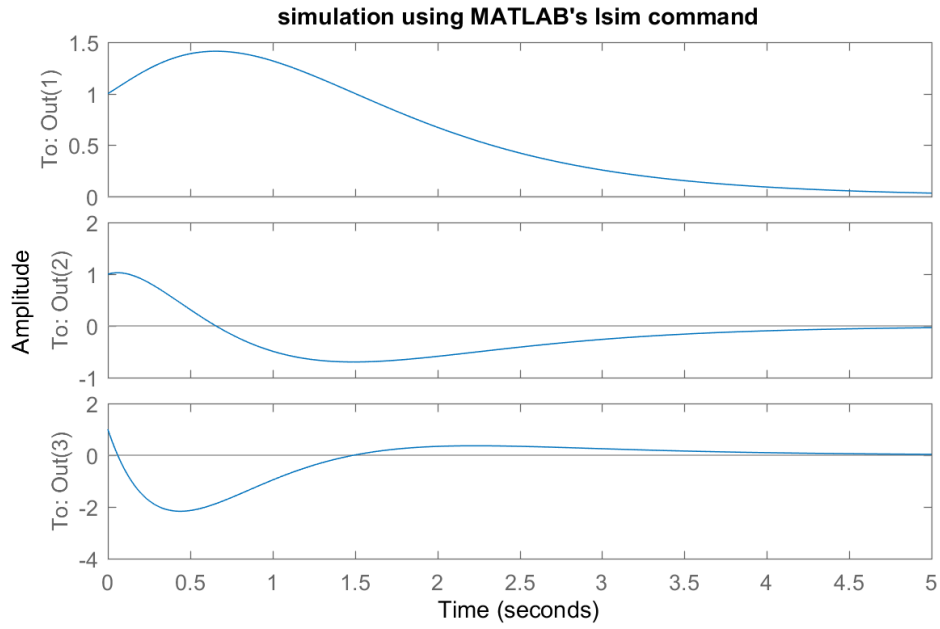
- c) Here is code that simulates using MATLAB's built-in `lsim` function, then manually using the matrix exponential (continuous-time solution), and finally using the discretized system. We can see that the three agree.

```
A = [0 1 0; 0 0 1; -5 -9 -5];
x0 = [1; 1; 1];

% simulation using LSIM command. To do this, we make B zero (no input), we
% make C the identity (return the state as output).
sys = ss(A,zeros(3,1),eye(3),0);
t = linspace(0,5,300);
u = 0*t; % zero input
figure(1)
lsim(sys,u,t,x0)
title('simulation using MATLAB''s lsim command')

% let's repeat the simulation, but this time computing the solution
% ourselves using the matrix exponential.
t = linspace(0,5,300);
x = zeros(3,300);
for k = 1:300
    x(:,k) = expm(A*t(k))*x0;
end
figure(2)
plot(t,x)
title('simulation using matrix exponential')
legend('x_1','x_2','x_3')

% now let's discretize and see what we get:
h = 0.2;
Ad = expm(A*h);
td = 0:h:5;
n = numel(td);
xd = zeros(3,n);
xd(:,1) = x0;
for j = 1:n-1
    xd(:,j+1) = Ad*xd(:,j);
end
figure(3)
plot(td,xd,'.-','MarkerSize',12)
title('simulation of discretized system')
legend('x_1','x_2','x_3')
```



**2. Dual state-space system.** Consider the MIMO system:

$$\begin{aligned}\dot{x}(t) &= Ax(t) + Bu(t) \\ y(t) &= Cx(t) + Du(t)\end{aligned}$$

Suppose the transfer function of this system (from  $u$  to  $y$ ) is given by  $H(s)$ . Now consider the *dual* system with state  $z(t)$  and state-space equations:

$$\begin{aligned}\dot{z}(t) &= -A^\top z(t) - C^\top v(t) \\ w(t) &= B^\top z(t) + D^\top v(t)\end{aligned}$$

Prove that the transfer function for this new system (from  $v$  to  $w$ ) is given by  $(H(-s))^\top$ .

**SOLUTION:** The transfer function for  $(A, B, C, D)$  is  $H(s) = C(sI - A)^{-1}B + D$ . So the transfer function for the dual system  $(-A^\top, -C^\top, B^\top, D^\top)$  is:  $B^\top(sI + A^\top)^{-1}(-C^\top) + D^\top$ . We can now check:

$$\begin{aligned}(H(-s))^\top &= (C(-sI - A)^{-1}B + D)^\top \\ &= B^\top(-sI - A)^{-\top}C^\top + D^\top \\ &= B^\top(sI + A^\top)^{-1}(-C^\top) + D^\top\end{aligned}$$

and this is the same as the previous expression.

**3. Diagonal form.** Consider the state-space equation  $\dot{x}(t) = Ax(t) + Bu(t)$  where

$$A = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -6 & -11 & -6 \end{bmatrix} \quad \text{and} \quad B = \begin{bmatrix} 0 \\ 1 \\ -3 \end{bmatrix}$$

- Use a state transformation to convert this system into diagonal canonical form.
- Is this system controllable? Explain your answer by analyzing your solution to part (a).
- Compute closed-form expressions for  $A^k$  and for  $e^{At}$ .

**SOLUTION:**

a) Based on the form of the  $A$  matrix, we know the characteristic polynomial is

$$\lambda^3 + 6\lambda^2 + 11\lambda + 6 = (\lambda + 1)(\lambda + 2)(\lambda + 3)$$

So the eigenvalues of  $A$  are  $\{-1, -2, -3\}$ . One possible diagonalization is:

$$\begin{bmatrix} 1 & 1 & 1 \\ -1 & -2 & -3 \\ 1 & 4 & 9 \end{bmatrix}^{-1} \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -6 & -11 & -6 \end{bmatrix} \begin{bmatrix} 1 & 1 & 1 \\ -1 & -2 & -3 \\ 1 & 4 & 9 \end{bmatrix} = \begin{bmatrix} -1 & 0 & 0 \\ 0 & -2 & 0 \\ 0 & 0 & -3 \end{bmatrix}$$

Computing the inverse, we have:

$$\begin{bmatrix} 1 & 1 & 1 \\ -1 & -2 & -3 \\ 1 & 4 & 9 \end{bmatrix}^{-1} = \frac{1}{2} \begin{bmatrix} 6 & 5 & 1 \\ -6 & -8 & -2 \\ 2 & 3 & 1 \end{bmatrix}$$

The diagonal form of the system is:  $(T^{-1}AT, T^{-1}B)$ , which we can compute to be:

$$\hat{A} = T^{-1}AT = \begin{bmatrix} -1 & 0 & 0 \\ 0 & -2 & 0 \\ 0 & 0 & -3 \end{bmatrix} \quad \text{and} \quad \hat{B} = T^{-1}B = \begin{bmatrix} 1 \\ -1 \\ 0 \end{bmatrix}$$

b) The system is not controllable. The controllability matrix (which we may compute in the transformed coordinates) is:

$$\begin{bmatrix} \hat{B} & \hat{A}\hat{B} & \hat{A}^2\hat{B} \end{bmatrix} = \begin{bmatrix} 1 & -1 & 1 \\ -1 & 2 & -4 \\ 0 & 0 & 0 \end{bmatrix}$$

and this matrix is clearly rank-deficient because it has a row of zeros.

c) We can also compute powers and exponentials using the diagonalized form:

$$A^k = \frac{1}{2} \begin{bmatrix} 6 & 5 & 1 \\ -6 & -8 & -2 \\ 2 & 3 & 1 \end{bmatrix} \begin{bmatrix} (-1)^k & 0 & 0 \\ 0 & (-2)^k & 0 \\ 0 & 0 & (-3)^k \end{bmatrix} \begin{bmatrix} 1 & 1 & 1 \\ -1 & -2 & -3 \\ 1 & 4 & 9 \end{bmatrix}$$

$$e^{At} = \frac{1}{2} \begin{bmatrix} 6 & 5 & 1 \\ -6 & -8 & -2 \\ 2 & 3 & 1 \end{bmatrix} \begin{bmatrix} e^{-t} & 0 & 0 \\ 0 & e^{-2t} & 0 \\ 0 & 0 & e^{-3t} \end{bmatrix} \begin{bmatrix} 1 & 1 & 1 \\ -1 & -2 & -3 \\ 1 & 4 & 9 \end{bmatrix}$$

**4. Linear time-varying problem.** Consider the time-varying linear dynamical system:

$$\begin{aligned} \dot{x}(t) &= A(t)x(t) + B(t)u(t) \\ y(t) &= C(t)x(t) + D(t)u(t) \end{aligned} \quad \text{with } x(0) = x_0 \quad (1)$$

a) Suppose we can find a matrix function  $P(t)$  with the property that

$$\dot{P}(t) = A(t)P(t) \quad \text{and } \det(P(t)) \neq 0 \quad \text{for all } t$$

A  $P(t)$  with this property is called a *fundamental matrix*. Note that in the case where  $A(t)$  is constant, we can use  $P(t) = e^{At}$ , as seen in class. Define  $\Phi(t, \tau) = P(t)P(\tau)^{-1}$ . Prove that the solution to (1) is given by:

$$y(t) = C(t)\Phi(t, 0)x_0 + \int_0^t C(t)\Phi(t, \tau)B(\tau)u(\tau) d\tau + D(t)u(t)$$

**Hint:** Use the transformation  $x(t) = P(t)z(t)$ .

b) Based on the result from part (a), it's enough to find a fundamental matrix  $P(t)$  and we automatically obtain a solution to the time-varying system (1). Consider the special case where  $x(t)$  is a scalar. In this case,  $A(t) = a(t)$  is also a scalar, and the fundamental matrix  $P(t) = p(t)$  is a scalar as well. Find  $p(t)$ .

c) Consider the sequence of functions  $\{M_k\}$  defined by:

$$\begin{aligned} M_0(t) &= I \\ M_{k+1}(t) &= I + \int_0^t A(\tau)M_k(\tau) d\tau \quad \text{for } k = 0, 1, \dots \end{aligned}$$

and let  $P(t) = \lim_{k \rightarrow \infty} M_k(t)$ . In other words,

$$\begin{aligned} P(t) &= I + \int_0^t A(\tau_1) d\tau_1 + \int_0^t A(\tau_1) \int_0^{\tau_1} A(\tau_2) d\tau_2 d\tau_1 \\ &\quad + \int_0^t A(\tau_1) \int_0^{\tau_1} A(\tau_2) \int_0^{\tau_2} A(\tau_3) d\tau_3 d\tau_2 d\tau_1 + \dots \end{aligned}$$

It turns out this is a well-defined limit (the sequence  $\{M_k\}$  converges uniformly). Prove that  $P(t)$  is a fundamental matrix.

### SOLUTION:

a) Using the transformation  $x(t) = P(t)z(t)$ , we have:

$$\dot{x}(t) = \frac{d}{dt}(P(t)z(t)) = \dot{P}(t)z(t) + P(t)\dot{z}(t) = A(t)P(t)z(t) + P(t)\dot{z}(t)$$

Substituting into the equation (1), we obtain the equations:

$$\begin{aligned} A(t)P(t)z(t) + P(t)\dot{z}(t) &= A(t)P(t)z(t) + B(t)u(t) \\ y(t) &= C(t)P(t)z(t) + D(t)u(t) \end{aligned}$$

The first equation has cancellations and simplifies to  $\dot{z}(t) = P(t)^{-1}B(t)u(t)$ . Therefore, we can integrate and conclude that:

$$z(t) = \int_0^t P(\tau)^{-1}B(\tau)u(\tau) d\tau$$

Substituting this into the second equation and letting  $\Phi(t, \tau) = P(t)P(\tau)^{-1}$  gives us the required result.

- b) If  $A(t) = a(t)$  is a scalar, the fundamental matrix is also a scalar  $P(t) = p(t)$ . We seek to solve the equation:

$$\frac{d}{dt} p(t) = a(t)p(t)$$

We saw in class how to solve this using an integrating factor. Another way is by writing it in differential form and integrating:

$$\begin{aligned} \frac{dp}{p} = a(t) dt &\implies \int_{p(0)}^{p(t)} \frac{1}{p} dp = \int_0^t a(\tau) d\tau \\ &\implies \log(p(t)) - \log(p(0)) = \int_0^t a(\tau) d\tau \\ &\implies p(t) = p(0)e^{\int_0^t a(\tau) d\tau} \end{aligned}$$

We can use any  $p(0) \neq 0$ . Ultimately, it doesn't matter since it will cancel out when we compute

$$\Phi(t_1, t_2) = \frac{p(t_1)}{p(t_2)} = e^{\int_{t_2}^{t_1} a(\tau) d\tau}$$

- c) The function  $P(t)$  is a series:  $P(t) = \sum_{k=0}^{\infty} P_k(t)$ . The terms in this series are:

$$\begin{aligned} P_0(t) &= I \\ P_1(t) &= \int_0^t A(\tau_1) d\tau \\ P_k(t) &= \int_0^t A(\tau_1) \int_0^{\tau_1} A(\tau_2) \int_0^{\tau_2} A(\tau_3) \cdots \int_0^{\tau_k} A(\tau_k) d\tau_k \cdots d\tau_3 d\tau_2 d\tau_1 \end{aligned}$$

Differentiating a general term with respect to  $t$ , we obtain  $\frac{d}{dt} P_0(t) = 0$  and for  $k \geq 1$ :

$$\begin{aligned} \frac{d}{dt} P_k(t) &= A(t) \int_0^t A(\tau_2) \int_0^{\tau_2} A(\tau_3) \cdots \int_0^{\tau_{k-1}} A(\tau_k) d\tau_k \cdots d\tau_3 d\tau_2 \\ &= A(t) \int_0^t A(\tau_1) \int_0^{\tau_1} A(\tau_2) \cdots \int_0^{\tau_{k-2}} A(\tau_{k-1}) d\tau_{k-1} \cdots d\tau_2 d\tau_1 \\ &= A(t)P_{k-1}(t) \end{aligned}$$

Where in the second step, we relabeled the dummy integration variables. Substituting this in to the series definition, we get:

$$\begin{aligned} \frac{d}{dt} P(t) &= \frac{d}{dt} \sum_{k=0}^{\infty} P_k(t) \\ &= \sum_{k=0}^{\infty} \frac{d}{dt} P_k(t) \\ &= \sum_{k=1}^{\infty} A(t)P_{k-1}(t) \\ &= A(t) \sum_{k=0}^{\infty} P_k(t) \\ &= A(t)P(t) \end{aligned}$$