

Generalized Necessary and Sufficient Robust Boundedness Results for Feedback Systems

Saman Cyrus*

Laurent Lessard[†]

Abstract

Classical sufficient conditions for ensuring the robust stability of a dynamical system in feedback with a nonlinearity include passivity, small gain, circle, and conicity theorems. We present a generalized version of these results for arbitrary semi-inner product spaces. Our result is purely algebraic, and holds even when the conventional discrete or continuous-time causal dynamical systems are replaced by general nonlinear relations, where there need not exist a notion of time. Our result clarifies when the sufficient conditions for robust stability are also necessary, and explains why stronger assumptions such as linearity and time-invariance are typically needed to prove necessity in the conventional dynamical systems setting.

1 Introduction

Robust stability of interconnected systems has been a topic of interest for over 75 years, dating back to the seminal works of Lur'e [14], Zames [30, 31], and Willems [28]. The standard input-output setup is illustrated in Fig. 1, where systems G and Φ are connected in feedback, and we seek conditions under which we can ensure the stability of the closed-loop map $(u_1, u_2) \rightarrow (y_1, y_2)$.

Robust stability results typically assume a known G is interconnected with some unknown, uncertain, or otherwise troublesome $\Phi \in \mathcal{C}_\Phi$, where \mathcal{C}_Φ is known. Then, if certain conditions on G and \mathcal{C}_Φ are met, we can ensure that the interconnection of Fig. 1 is stable.

There are many robust stability results in the literature: passivity theory, the small-gain theorem, the circle criterion, graph separation, conic sector theorems, multiplier theory, dissipativity theory, and integral quadratic constraints.¹

The reason for the wide variety of robust stability results is that different assumptions can be made about G and \mathcal{C}_Φ . For example, G and Φ are typically causal operators on an extended space of time-domain signals such as L_{2e} or ℓ_{2e} . Additionally, G or Φ may be restricted to be linear, time-invariant, or static. Finally, some results are stated as sufficient conditions while others are both necessary and sufficient.

In spite of their diversity, robust stability results are typically proven using the same elementary properties of inner product spaces. A natural question to ask, which forms the basis of our present work, is whether the multitude of existing results can be viewed as consequences of a purely algebraic result. We answer in the affirmative.

*Department of Electrical and Computer Engineering, University of Wisconsin–Madison. cyrus2@wisc.edu

[†]Department of Mechanical and Industrial Engineering, Northeastern University. l.1.1essard@northeastern.edu

¹Detailed references can be found in Section 1.1 and Table 1.



Figure 1: Feedback interconnection of systems G and Φ .

Main contribution: In Section 2, we present Theorem 1, a robust boundedness result involving interconnected *relations* over a general *semi-inner product space*. Theorem 1 distills the vast literature on robust stability into a simple and purely algebraic result.

In Section 3, we specialize Theorem 1 to L_{2e} and ℓ_{2e} spaces, which reveals the connections between the algebraic version of the result and notions of *well-posedness*, *causality*, and *stability*. We also explain why stronger assumptions, such as linearity and time-invariance of G , are often required in order to achieve both sufficiency and necessity.

1.1 Related work

In Table 1, we provide a summary of existing robust stability results. In the “Direction” column, we distinguish between *sufficient-only* results (\implies) and *necessary-and-sufficient* results (\iff).

Sufficient results. Classical sufficient results include the passivity, small-gain, and circle theorems. These results are mutually related via a loop-shifting transformation [1], and were generalized to *conic sectors* [2, 30, 31].

Beyond conic sector constraints, graph separation [20, 24] allows for nonlinear constraints, while multiplier theory [9], dissipativity [28], and integral quadratic constraints (IQCs) [17, 19, 26] allow for dynamic or time-varying constraints. There have also been several works discussing how these various frameworks are related [5, 10, 21]. In Table 1, we distinguish between *static* constraints (the focus of the present work), and more general *dynamic* constraints, which include multipliers, dissipativity theory, and IQCs.

Necessary and sufficient results. When Φ is assumed to be memoryless (but still possibly time-varying), the classical passivity, small-gain, and circle theorems are only sufficient for robust stability [4, 16].

Finding a robust stability condition that is both sufficient and necessary requires stronger assumptions. The set \mathcal{C}_Φ must be broadened to allow dynamic nonlinearities, and we must typically assume that G is linear and time-invariant (LTI). For example, the passivity and small gain results of Vidyasagar [27, §6.6(112,126)] and Khong et al. [13, Thm. 3] assume G is LTI. The small-gain result of Zhou et al. [32, Thm. 9.1] and the converse IQC result of Khong et al. [12] make the stronger assumption that *both* G and Φ are LTI. Finally, Shamma’s small-gain result [22, Thm. 3.2] holds when both G and Φ are nonlinear and time-varying, but requires a *fading memory* assumption, which allows the system response to be approximated by that of a linear system.

1.2 Notation

Preliminaries. The set \mathbb{F} refers to the field of real or complex numbers. The complex conjugate of $x \in \mathbb{F}$ is \bar{x} and the conjugate transpose of $X \in \mathbb{F}^{m \times n}$ is X^* . We use \preceq , \prec , \succ , \succeq to denote the (semi)definite partial ordering in $\mathbb{F}^{n \times n}$.

Table 1: Literature Review of robust stability results involving two interconnected systems (see Fig. 1). The first group of rows are sufficient-only results (\implies). The second group are necessary-and-sufficient (\iff). For constraints on G and $\Phi \in \mathcal{C}_\Phi$ we denote linear (L), nonlinear (N), time-varying (TV), time-invariant (TI), static (S), and fading-memory (F). For example, “LTI” indicates linear and time-invariant. Symbols L_{2e} and ℓ_{2e} denote extended spaces (L_{2e} for both) and s.i.p.s. denotes a semi-inner product space. In extended spaces, G and Φ are constrained to be causal. There is no such requirement for s.i.p.s., since there need not exist a notion of time. The final column indicates whether the converse proof direction (\impliedby), if applicable, explicitly constructs a worst-case Φ when the conditions on G are violated. The present work is restricted to static constraints.

Reference	Constraint	Result Type	Space	G	\mathcal{C}_Φ	Direction	Constructive
Vidyasagar [27, §6.6.(1,58)]	static	passivity & small gain	\mathcal{L}_{2e}	N	N	\implies	
Zames [30, Thm. 1–3]	static	conic	\mathcal{L}_{2e}	N	N	\implies	
Bridgeman & Forbes [3]	static	conic	\mathcal{L}_{2e}	N	N	\implies	
Zames [31, §3–4]	static	circle & multipliers	L_{2e}	LTI	NS	\implies	
Desoer & Vidyasagar [9]	dynamic	multipliers	\mathcal{L}_{2e}	N	N	\implies	
Teel et al. [24]	static	graph separation	\mathcal{L}_{2e}	N	N	\implies	
Willems [28]	dynamic	dissipativity	\mathcal{L}_{2e}	N	N	\implies	
Pfifer & Seiler [19]	dynamic	dissipativity	L_{2e}	LTI	N	\implies	
Megretski & Rantzer [17]	dynamic	IQC	L_{2e}	LTI	NS	\implies^\dagger	
Vidyasagar [27, §6.6.(112,126)]	static	small gain & circle	L_{2e}	LTI	N	\iff	Yes
Khong & van der Schaft [13, Thm. 3]	static	passivity & small gain	L_{2e}	LTI	LTV	\iff	No
Zhou et al. [32, Thm. 9.1]	static	small gain	L_{2e}	LTI	LTI	\iff	Yes
Khong & Kao [12, Thm. 1]	dynamic	IQC	L_{2e}	LTI	LTI	\iff	Yes
Shamma [22, Thm. 3.2]	static	small gain	ℓ_{2e}	NF	NF	\iff	Yes
Cyrus & Lessard [8]	static	conic	s.i.p.s.	L	N	\iff	No
Present work	static	conic	s.i.p.s.	N	N	\iff	Yes

[†] The authors in [17] mention that their sufficient condition for robust stability is also necessary in the sense that a result in the spirit of Lemma 1 holds via a suitable application of the S-lemma [18].

Semi-inner products. A *semi-inner product space* is a vector space \mathcal{V} over a field \mathbb{F} equipped with a semi-inner product² $\langle \cdot, \cdot \rangle$, which is an inner product whose associated norm is a seminorm. In other words, $\|x\| := \sqrt{\langle x, x \rangle} \geq 0$ for all $x \in \mathcal{V}$, but $\|x\| = 0$ does not imply that $x = 0$.

Relations. A *relation* R on \mathcal{V} is a subset of the product space $R \subseteq \mathcal{V} \times \mathcal{V}$. We write $\mathcal{R}(\mathcal{V})$ to denote the set of all relations on \mathcal{V} . The **domain** of R is $\text{dom}(R) := \{x \in \mathcal{V} \mid (x, y) \in R \text{ for some } y \in \mathcal{V}\}$. For any $x \in \text{dom}(R)$, we write Rx to denote any $y \in \mathcal{V}$ such that $(x, y) \in R$.

We define \mathcal{V}^2 as augmented vectors $\begin{pmatrix} u_1 \\ u_2 \end{pmatrix}$ where $u_1, u_2 \in \mathcal{V}$. We overload matrix multiplication in \mathcal{V}^2 ; for any $\xi, \zeta \in \mathcal{V}^2$ and any matrix $N \in \mathbb{F}^{2 \times 2}$,

$$N\xi = \begin{bmatrix} N_{11} & N_{12} \\ N_{21} & N_{22} \end{bmatrix} \begin{bmatrix} \xi_1 \\ \xi_2 \end{bmatrix} := \begin{bmatrix} N_{11}\xi_1 + N_{12}\xi_2 \\ N_{21}\xi_1 + N_{22}\xi_2 \end{bmatrix} \in \mathcal{V}^2.$$

Likewise, inner products in \mathcal{V}^2 have the interpretation

$$\langle \xi, \zeta \rangle = \left\langle \begin{bmatrix} \xi_1 \\ \xi_2 \end{bmatrix}, \begin{bmatrix} \zeta_1 \\ \zeta_2 \end{bmatrix} \right\rangle := \langle \xi_1, \zeta_1 \rangle + \langle \xi_2, \zeta_2 \rangle.$$

We omit subscripts when referring to many of the u_i, y_i, e_i from Fig. 1 at once. For example, (u, y, e) is shorthand for $(u_1, u_2, y_1, y_2, e_1, e_2)$. We also define the following relations, which characterize pairs of consistent signals.

$$\begin{aligned} R_{uy} &:= \{(u, y) \in \mathcal{V}^2 \times \mathcal{V}^2 \mid (1) \text{ holds for some } e \in \mathcal{V}^2\}, \\ R_{ue} &:= \{(u, e) \in \mathcal{V}^2 \times \mathcal{V}^2 \mid (1) \text{ holds for some } y \in \mathcal{V}^2\}. \end{aligned}$$

²We use the convention that a semi-inner product is linear in its second argument, so $\langle x, ay + bz \rangle = a\langle x, y \rangle + b\langle x, z \rangle$ for all $x, y, z \in \mathcal{V}$ and $a, b \in \mathbb{F}$. Also, $\langle x, y \rangle = \overline{\langle y, x \rangle}$.

2 Results for semi-inner product spaces

Our main result is a robust boundedness theorem defined over a general semi-inner product space. We consider the setup of Fig. 1, where $G \in \mathcal{R}(\mathcal{V})$ and $\Phi \in \mathcal{C}_\Phi \subseteq \mathcal{R}(\mathcal{V})$ are (possibly nonlinear) relations.

Theorem 1. *Let \mathcal{V} be a semi-inner product space and let $M = M^* \in \mathbb{F}^{2 \times 2}$. Suppose $G \in \mathcal{R}(\mathcal{V})$ and $\mathcal{C}_\Phi \subseteq \mathcal{R}(\mathcal{V})$. Consider the three following statements.*

(i) *There exists $N = N^* \in \mathbb{F}^{2 \times 2}$ satisfying $M + N \prec 0$ such that the following property of G holds.*

$$\left\langle \begin{bmatrix} G\xi \\ \xi \end{bmatrix}, N \begin{bmatrix} G\xi \\ \xi \end{bmatrix} \right\rangle \geq 0 \quad \text{for all } \xi \in \text{dom}(G). \quad (3)$$

(ii) *There exists $\gamma > 0$ such that for all (u, y, e) , if*

$$\left\langle \begin{bmatrix} e_2 \\ y_2 \end{bmatrix}, M \begin{bmatrix} e_2 \\ y_2 \end{bmatrix} \right\rangle \geq 0 \quad (4)$$

and (1a), (1c), (1d) are satisfied, then $\|y\| \leq \gamma\|u\|$.

(iii) *There exists $\gamma > 0$ such that for all $\Phi \in \mathcal{C}_\Phi$, if*

$$\left\langle \begin{bmatrix} \xi \\ \Phi\xi \end{bmatrix}, M \begin{bmatrix} \xi \\ \Phi\xi \end{bmatrix} \right\rangle \geq 0 \quad \text{for all } \xi \in \text{dom}(\Phi), \quad (5)$$

then for all $(u, y) \in R_{uy}$, the following bound holds

$$\|y\| \leq \gamma\|u\|. \quad (6)$$

The following equivalences hold:

- *Graph separation: (i) \iff (ii).*
- *Interpolation: (ii) \implies (iii).*

A pedagogical benefit of Theorem 1 is that it splits the robustness result into a *graph separation* statement that concerns G and an *interpolation* statement that concerns \mathcal{C}_Φ .

The result (i) \iff (ii) relates boundedness G in (3) to boundedness of the closed-loop map when Φ is replaced by the inequality (4). This graph separation result holds for arbitrary G (any nonlinear relation), and does not depend on Φ or \mathcal{C}_Φ .

The result (ii) \implies (iii) relates the inequality (5) satisfied by Φ to the inequality (4) satisfied by the inputs and outputs of Φ . Whether or not the converse holds depends on whether the set \mathcal{C}_Φ is rich enough to allow interpolation. In other words, given the signals e_2 and y_2 satisfying (4), does there necessarily exist a $\Phi \in \mathcal{C}_\Phi$ such that $y_2 = \Phi e_2$?

Theorem 1 is *sufficient* for robust boundedness because it proves (i) \implies (iii). In Section 2.3, we show that with suitable assumptions about G and \mathcal{C}_Φ , we can satisfy the interpolation requirement and therefore make the result necessary as well.

Since Theorem 1 is expressed using a general semi-inner product space, it holds even when G is not a dynamical system but rather a general nonlinear relation. So there need not exist a notion of time. We make a few additional remarks.

Remark 1. Equation (6) can be stated in terms of (u, e) instead of (u, y) . Specifically, (6) holds for all $(u, y) \in R_{uy}$ if and only if there exists some $\bar{\gamma} > 0$ such that $\|e\| \leq \bar{\gamma}\|u\|$ holds for all $(u, e) \in R_{ue}$.

Remark 2. In Item (i), we can equivalently replace N by $-M - \varepsilon I$ and modify the statement preceding (3) to: “There exists some $\varepsilon > 0$ such that G satisfies (3)”. We chose the form with M and N for aesthetic reasons.

Remark 3. Theorem 1 can be generalized to $G \in \mathcal{R}(\mathcal{V}^n, \mathcal{V}^m)$ (the set of relations on $\mathcal{V}^n \times \mathcal{V}^m$) and $\Phi \in \mathcal{C}_\Phi \subseteq \mathcal{R}(\mathcal{V}^m, \mathcal{V}^n)$. Here, $M, N \in \mathbb{F}^{(m+n) \times (m+n)}$ would be block 2×2 matrices.

2.1 Proof of sufficiency for Theorem 1

We begin by showing that (i) \implies (ii) \implies (iii). This proof is similar to [15, Thm. 1]. Pick any (u, y, e) such that (1a), (1c), (1d), and (4) are satisfied. Let $\xi = e_1$ in (3). Using (1) to eliminate e_1, e_2 , Equations (3) and (4) become: $\langle \begin{bmatrix} y_1 \\ u_1+y_2 \end{bmatrix}, N \begin{bmatrix} y_1 \\ u_1+y_2 \end{bmatrix} \rangle \geq 0$ and $\langle \begin{bmatrix} u_2+y_1 \\ y_2 \end{bmatrix}, M \begin{bmatrix} u_2+y_1 \\ y_2 \end{bmatrix} \rangle \geq 0$. Summing these two inequalities and collecting terms, we obtain

$$\left\langle \begin{bmatrix} y_1 \\ y_2 \end{bmatrix}, (M + N) \begin{bmatrix} y_1 \\ y_2 \end{bmatrix} \right\rangle + 2 \left\langle \begin{bmatrix} y_1 \\ y_2 \end{bmatrix}, \begin{bmatrix} N_{12} & M_{11} \\ N_{22} & M_{21} \end{bmatrix} \begin{bmatrix} u_1 \\ u_2 \end{bmatrix} \right\rangle + \left\langle \begin{bmatrix} u_1 \\ u_2 \end{bmatrix}, \begin{bmatrix} N_{22} & 0 \\ 0 & M_{11} \end{bmatrix} \begin{bmatrix} u_1 \\ u_2 \end{bmatrix} \right\rangle \geq 0.$$

Since $M + N \prec 0$ by assumption, There exists $\eta > 0$ such that $M + N \preceq -\eta I$. Applying this inequality together with Cauchy–Schwarz³, we get $-\eta\|y\|^2 + 2r\|y\|\|u\| + q\|u\|^2 \geq 0$, where $r := \left\| \begin{bmatrix} N_{12} & M_{11} \\ N_{22} & M_{21} \end{bmatrix} \right\|$ and $q := \left\| \begin{bmatrix} N_{22} & 0 \\ 0 & M_{11} \end{bmatrix} \right\|$ are standard spectral norms. Dividing by η and completing the square, we can rewrite the last inequality as $(\|y\| - \frac{r}{\eta}\|u\|)^2 \leq \frac{r^2 + \eta q}{\eta^2}\|u\|^2$, which can be rearranged to establish (ii) with $\gamma = \frac{1}{\eta}(r + \sqrt{r^2 + \eta q})$.

To prove (iii), consider some $\Phi \in \mathcal{C}_\Phi$ for which (5) holds. Next, pick $(u, y) \in R_{uy}$ so that there exists (u, y, e) satisfying (1). In particular, (1b) holds, so setting $\xi = e_2$ in (5), we obtain (4) and the rest of the proof is the same as above. \blacksquare

2.2 Necessity of graph separation in Theorem 1

A popular approach for proving (i) \iff (ii) is to use a lossless S-lemma as in [13, Thm. 3] and [17]. However, the S-lemma [18, 29] comes with a drawback: the set of signals (u, y, e) that satisfy the loop equations (1a), (1c), (1d) must be a *subspace*, which requires for example that G be linear. If we assume G is linear, we can prove (i) \iff (ii) by adapting the S-lemma for inner product spaces due to Hestenes [11, Thm. 7.1, p. 354] and using a technique similar to that used in [13]. Details of this approach may be found in [7, 8].

The linearity assumption on G can be dropped entirely if we adopt a different proof approach. To this effect, we will prove the contrapositive $\neg(\text{i}) \implies \neg(\text{ii})$ by directly constructing signals (y, u, e) that violate the boundedness condition when (i) fails to hold. Unlike the S-lemma, this approach does not require linearity of G and has the benefit of being constructive, so it produces *worst-case* signals (u, y, e) .

³A proof of the Cauchy–Schwarz inequality for general semi-inner product spaces may be found in [6, §1.4].

Lemma 1 (worst-case signals). *Consider the setting of Theorem 1. Suppose that for any N satisfying $M + N \prec 0$, there exists $\xi \in \text{dom}(G)$ such that $\left\langle \begin{bmatrix} G\xi \\ \xi \end{bmatrix}, N \begin{bmatrix} G\xi \\ \xi \end{bmatrix} \right\rangle < 0$. Then, for all $\gamma > 0$, there exists (u, y, e) such that:*

1. Equations (1a), (1c), and (1d) hold.
2. $\left\langle \begin{bmatrix} e_2 \\ y_2 \end{bmatrix}, M \begin{bmatrix} e_2 \\ y_2 \end{bmatrix} \right\rangle \geq 0$.
3. $\|y\| > \gamma\|u\|$.

Proof. See Appendix A.1 for a detailed proof. ■

The implication (i) \iff (ii) of Theorem 1 now directly follows from Lemma 1.

2.3 Necessity of interpolation in Theorem 1

The implications (i) \iff (ii) \implies (iii) of Theorem 1 hold with great generality. However, the missing implication (ii) \iff (iii) does not hold in general, because it depends on the choice of G and \mathcal{C}_Φ . If \mathcal{C}_Φ is insufficiently expressive, there may not exist a $\Phi \in \mathcal{C}_\Phi$ that interpolates the closed-loop signals found in (ii). We now explore some special cases for which the missing implication holds; in other words, there exists a Φ that interpolates the closed-loop signals.

Definition 1. *We say that a pair (G, \mathcal{C}_Φ) is interpolable if such a choice implies that (ii) \iff (iii) in Theorem 1.*

We now describe simple scenarios in which interpolability is guaranteed for the general semi-inner product setting. First, we make the trivial observation that if Φ is unconstrained, interpolation is always possible.

Proposition 1 (unconstrained case). *If $\mathcal{C}_\Phi = \mathcal{R}(\mathcal{V})$, then the pair (G, \mathcal{C}_Φ) is interpolable for any $G \in \mathcal{R}(\mathcal{V})$.*

Proof. If $\mathcal{C}_\Phi = \mathcal{R}(\mathcal{V})$, then (4) and (5) are equivalent, as we can choose the singleton relation $\Phi = \{(e_2, y_2)\}$. ■

Proposition 1 is not particularly satisfying because it requires the use of a singleton relation Φ . A more interesting case is when we require that $\text{dom}(\Phi) = \mathcal{V}$.⁴ Our second result states that interpolability holds for the set of *linear* relations.

Definition 2 (linear relation). *Let \mathcal{V} be a semi-inner product space over a field \mathbb{F} . Let $x_1, x_2, y_1, y_2 \in \mathcal{V}$ and $\alpha_1, \alpha_2 \in \mathbb{F}$. A relation $R \in \mathcal{R}(\mathcal{V})$ is linear if for all $(x_1, y_1) \in R$ and $(x_2, y_2) \in R$, we have $(\alpha_1 x_1 + \alpha_2 x_2, \alpha_1 y_1 + \alpha_2 y_2) \in R$. We let $\mathcal{L}(\mathcal{V}) \subseteq \mathcal{R}(\mathcal{V})$ denote the set of all linear relations.*

Theorem 2 (linear case). *If $\mathcal{C}_\Phi = \mathcal{L}(\mathcal{V})$, then the pair (G, \mathcal{C}_Φ) is interpolable for any $G \in \mathcal{R}(\mathcal{V})$.*

Proof. We explicitly construct a worst-case $\Phi \in \mathcal{L}(\mathcal{V})$. See Appendix A.2 for a detailed proof. ■

Proposition 1 and Theorem 2 both provide conditions that ensure necessity of Theorem 1. In both cases, there are no constraints on G ; it could be nonlinear, for example.

⁴Relations $\Phi \in \mathcal{R}(\mathcal{V})$ that satisfy $\text{dom}(\Phi) = \mathcal{V}$ are known as *serial* or *left-total*. They are also called *multi-valued functions*.

3 Specialization to extended spaces

The most common application of robust stability is when (y, u, e) are time-domain signals belonging to an extended space such as L_{2e} or ℓ_{2e} [32]. This forces us to deal with well-posedness, causality, and stability.

Well-posedness. Assuming G and Φ are relations, as we do in Theorem 1, is not unprecedented in the literature [13,20,25,27,30]. This ensures the closed-loop relations R_{uy} and R_{ue} are always well-defined, but they may be *empty*. When G and Φ are assumed to be operators instead of relations, then well-posedness must either be assumed or proved. Specifically, we need an assurance of the existence and uniqueness of solutions e and y for all choices of u .

Causality. When working in extended spaces such as \mathcal{L}_{2e} , a common assumption is that G and Φ are causal operators [9,17,23,27,31,32]. Then, a useful fact is that a well-posed interconnection of causal maps is causal [25, Prop. 1.2.14], so the closed-loop map will be causal.

Stability. The goal when working with time-domain signals is typically to prove stability. With Theorem 1, we prove boundedness of the closed-loop map, i.e., $\|y\| \leq \gamma\|u\|$, and therefore input-output stability.

To specialize Theorem 1 to extended spaces, set $\mathcal{V} = \mathcal{L}_{2e}$ and use the semi-inner product $\langle \cdot, \cdot \rangle_T$ defined by projecting both signals onto $[0, T]$ and applying the \mathcal{L}_{2e} inner product. Then, use the fact that if H is a causal map, $\|Hx\| \leq \gamma\|x\|$ for all $x \in \mathcal{L}_2$ if and only if $\|Hx\|_T \leq \gamma\|x\|_T$ for all x and T (see, for example, [27, Lem. 6.2.11]).

Different choices of the matrices M and N allow the representation of different cones. For example, we can represent different flavors of passivity (input-strict, output-strict, extended), small-gain results, the circle criterion, and other conic sectors that allow G or Φ to be unbounded/unstable.

To illustrate these various transformations, consider for example the classical passivity result by Vidyasagar, which is a sufficient-only result, and may be found in [27, Thm. 6.7.43].

Theorem 3 (Vidyasagar). *Consider the system*

$$\begin{cases} e_1 = u_1 - y_2, & y_1 = Ge_1 \\ e_2 = u_2 + y_1, & y_2 = \Phi e_2 \end{cases}$$

Suppose there exist constants $\varepsilon_1, \varepsilon_2, \delta_1, \delta_2$ such that for all $\xi \in \ell_{2e}$ and for all $T \geq 0$

$$\langle \xi, G\xi \rangle_T \geq \varepsilon_1 \|\xi\|_T^2 + \delta_1 \|G\xi\|_T^2, \tag{7a}$$

$$\langle \xi, \Phi\xi \rangle_T \geq \varepsilon_2 \|\xi\|_T^2 + \delta_2 \|\Phi\xi\|_T^2. \tag{7b}$$

Then the system is ℓ_2 -stable if $\delta_1 + \varepsilon_2 > 0$ and $\delta_2 + \varepsilon_1 > 0$.

Theorem 3 uses a negative sign convention and is expressed in discrete time. To match 1, let $\Phi \mapsto -\Phi$ in Theorem 3 and compare (3) and (5) to (7), which yields

$$N = \begin{bmatrix} -\delta_1 & \frac{1}{2} \\ \frac{1}{2} & -\varepsilon_1 \end{bmatrix} \quad \text{and} \quad M = \begin{bmatrix} -\varepsilon_2 & -\frac{1}{2} \\ -\frac{1}{2} & -\delta_2 \end{bmatrix}.$$

In Theorem 1, we require $M + N \prec 0$; thus $\delta_1 + \varepsilon_2 > 0$ and $\delta_2 + \varepsilon_1 > 0$, which recovers Theorem 3. A similar approach can be used to recover all the results from Table 1 involving static constraints. For a detailed proof, see [7].

Unlike Theorem 1, Theorem 2 does not specialize as nicely to extended spaces. In particular, the construction of a worst-case Φ from Lemma 2 will not, in general, be causal. In order to achieve interpolability, one must typically make additional assumptions, such as G and Φ being linear and time-invariant (see Table 1). In such cases, a worst-case Φ can be chosen as a static gain cascaded with a time delay [27, §6.6.(112,126)].

4 Conclusion

We studied robust stability results involving a plant G connected with a nonlinearity Φ belonging to a conic sector, e.g. passivity, small-gain, circle criterion, conicity, or extended conicity. Our goal was to distill the vast literature on this topic and state the most general and unified results possible.

Looking beyond the scope of this paper, it would be interesting to see if our semi-inner product framework could be used to recover results involving dynamic constraints (dissipativity, multiplier theory, integral quadratic constraints).

5 Acknowledgments

The authors would like to thank R. Boczar, L. Bridgeman, R. Brockett, S. Z. Khong, A. Packard, A. Rantzer, P. Seiler, A. van der Schaft, B. Van Scoy, and M. Vidyasagar for helpful discussions and comments through various stages of this work.

A Proofs for semi-inner product spaces

A.1 Proof of Lemma 1

Proof. The result is trivial or vacuous if M is semidefinite, so we will assume M is indefinite, writing

$$M = P^*JP, \quad (8)$$

where $J := \text{diag}(1, -1)$ and $P \in \mathbb{F}^{2 \times 2}$ is invertible. Pick any $0 < \varepsilon < 1$ and let $N = -M - \varepsilon P^*P$. By assumption, we can choose some $\xi \in \text{dom}(G)$ such that

$$\left\langle \begin{bmatrix} G\xi \\ \xi \end{bmatrix}, N \begin{bmatrix} G\xi \\ \xi \end{bmatrix} \right\rangle < 0. \quad (9)$$

Now pick $e_1 = \xi$, $y_1 = G\xi$, and

$$\begin{bmatrix} e_2 \\ y_2 \end{bmatrix} = \begin{bmatrix} y_1 \\ e_1 \end{bmatrix} + \begin{bmatrix} u_2 \\ -u_1 \end{bmatrix}, \text{ with } \begin{bmatrix} u_2 \\ -u_1 \end{bmatrix} = \varepsilon P^{-1}JP \begin{bmatrix} y_1 \\ e_1 \end{bmatrix}.$$

By construction, this choice satisfies Item 1 of Lemma 1. Substituting our choice of N into (9), we obtain

$$\left\langle \begin{bmatrix} y_1 \\ e_1 \end{bmatrix}, M \begin{bmatrix} y_1 \\ e_1 \end{bmatrix} \right\rangle > -\varepsilon \left\| P \begin{bmatrix} y_1 \\ e_1 \end{bmatrix} \right\|^2. \quad (10)$$

Substituting the definitions of u_1 , u_2 , e_2 , y_2 , M in terms of y_1 , e_1 , ε , P , J , and using the inequality (10), we have

$$\begin{aligned} \left\langle \begin{bmatrix} e_2 \\ y_2 \end{bmatrix}, M \begin{bmatrix} e_2 \\ y_2 \end{bmatrix} \right\rangle &= \left\langle (I + \varepsilon P^{-1}JP) \begin{bmatrix} y_1 \\ e_1 \end{bmatrix}, M(I + \varepsilon P^{-1}JP) \begin{bmatrix} y_1 \\ e_1 \end{bmatrix} \right\rangle \\ &= \left\langle (I + \varepsilon J)P \begin{bmatrix} y_1 \\ e_1 \end{bmatrix}, P^{-*}MP^{-1}(I + \varepsilon J)P \begin{bmatrix} y_1 \\ e_1 \end{bmatrix} \right\rangle \\ &\stackrel{(8)}{=} \left\langle (I + \varepsilon J)P \begin{bmatrix} y_1 \\ e_1 \end{bmatrix}, J(I + \varepsilon J)P \begin{bmatrix} y_1 \\ e_1 \end{bmatrix} \right\rangle \\ &= \left\langle P \begin{bmatrix} y_1 \\ e_1 \end{bmatrix}, ((1 + \varepsilon^2)J + 2\varepsilon I)P \begin{bmatrix} y_1 \\ e_1 \end{bmatrix} \right\rangle \\ &\stackrel{(8)}{=} (1 + \varepsilon^2) \left\langle \begin{bmatrix} y_1 \\ e_1 \end{bmatrix}, M \begin{bmatrix} y_1 \\ e_1 \end{bmatrix} \right\rangle + 2\varepsilon \left\| P \begin{bmatrix} y_1 \\ e_1 \end{bmatrix} \right\|^2 \\ &\stackrel{(10)}{>} \varepsilon(1 - \varepsilon^2) \left\| P \begin{bmatrix} y_1 \\ e_1 \end{bmatrix} \right\|^2 \geq 0, \end{aligned}$$

which verifies Item 2 of Lemma 1. Finally, we have:

$$\left\| \begin{bmatrix} u_1 \\ u_2 \end{bmatrix} \right\| = \left\| \begin{bmatrix} u_2 \\ -u_1 \end{bmatrix} \right\| = \left\| \varepsilon P^{-1}JP \begin{bmatrix} y_1 \\ e_1 \end{bmatrix} \right\| \leq \varepsilon \kappa \left\| \begin{bmatrix} y_1 \\ e_1 \end{bmatrix} \right\|,$$

where $\kappa := \|P^{-1}JP\| > 0$. Applying the triangle inequality,

$$\|u\| = \left\| \begin{bmatrix} u_1 \\ u_2 \end{bmatrix} \right\| \leq \varepsilon \kappa \left\| \begin{bmatrix} y_1 \\ e_1 \end{bmatrix} \right\| = \varepsilon \kappa \left\| \begin{bmatrix} y_1 \\ y_2 + u_1 \end{bmatrix} \right\| \leq \varepsilon \kappa \left(\left\| \begin{bmatrix} y_1 \\ y_2 \end{bmatrix} \right\| + \left\| \begin{bmatrix} 0 \\ u_1 \end{bmatrix} \right\| \right) \leq \varepsilon \kappa (\|y\| + \|u\|).$$

Rearranging, we obtain $\|y\| \geq \frac{1 - \varepsilon \kappa}{\varepsilon \kappa} \|u\|$. Since ε can be chosen arbitrarily small, we can make the bound γ arbitrarily large in Item 3 of Lemma 1, thus completing the proof. \blacksquare

A.2 Proof of Theorem 2

We begin by proving that a pair of points satisfying a quadratic constraint can be extended to a linear relation that satisfies the quadratic constraint everywhere.

Lemma 2 (extension lemma). *Let \mathcal{V} be a semi-inner product space and let $M = M^* \in \mathbb{F}^{2 \times 2}$. Suppose $e, y \in \mathcal{V}$ satisfy*

$$\left\langle \begin{bmatrix} e \\ y \end{bmatrix}, M \begin{bmatrix} e \\ y \end{bmatrix} \right\rangle \geq 0. \quad (11)$$

There exists $\Phi \in \mathcal{L}(\mathcal{V})$ such that:

1. $(e, y) \in \Phi$.
2. $\left\langle \begin{bmatrix} x \\ \Phi x \end{bmatrix}, M \begin{bmatrix} x \\ \Phi x \end{bmatrix} \right\rangle \geq 0$ for all $x \in \text{dom}(\Phi)$.

Moreover, if $\|e\| > 0$, we can construct Φ that is a linear function, with $\text{dom}(\Phi) = \mathcal{V}$.

Using Lemma 2, we can prove Theorem 2 by contradiction. Indeed, if Item (ii) of Theorem 1 fails, then for any $\gamma > 0$, there exist $e_2, y_2 \in \mathcal{V}$ such that (4) and (1a), (1c), (1d) hold, with $\|y\| > \gamma\|u\|$. Applying Lemma 2 to the pair (e_2, y_2) , we can produce $\Phi \in \mathcal{L}(\mathcal{V}) \subseteq \mathcal{C}_\Phi$ such that (5) holds, and thus (1b) holds, $(u, y) \in R_{uy}$, and therefore Item (iii) of Theorem 1 fails, as required. All that remains is to prove Lemma 2.

Proof. We begin by considering some special cases.

Special case with $\|e\| = 0$. Here, $\langle e, y \rangle = 0$ by Cauchy–Schwarz. If $\|y\| = 0$, define $\Phi := \{(z, x) \mid \|z\| = \|x\| = 0\}$. This is a degenerate case. If $\|y\| > 0$ instead, we have by assumption that $M_{22}\|y\|^2 = \left\langle \begin{bmatrix} e \\ y \end{bmatrix}, M \begin{bmatrix} e \\ y \end{bmatrix} \right\rangle \geq 0$. Therefore, $M_{22} \geq 0$. Define $\Phi = \{(z, x) \mid \|z\| = 0\}$. Roughly, Φ is the linear relation whose graph is a *vertical line*.

Special case with $\|e\| > 0$ and $\|y\| = 0$. As in the previous case, we must have $\langle e, y \rangle = 0$. By assumption, $M_{11}\|e\|^2 = \left\langle \begin{bmatrix} e \\ y \end{bmatrix}, M \begin{bmatrix} e \\ y \end{bmatrix} \right\rangle \geq 0$. So, $M_{11} \geq 0$. Let $\Phi x := 0$, so $\left\langle \begin{bmatrix} x \\ \Phi x \end{bmatrix}, M \begin{bmatrix} x \\ \Phi x \end{bmatrix} \right\rangle = M_{11}\|x\|^2 \geq 0$ for all $x \in \mathcal{V}$. Henceforth, we will assume that $\|e\| > 0$ and $\|y\| > 0$. Define the normalized vectors $\hat{e} := \frac{e}{\|e\|}$ and $\hat{y} := \frac{y}{\|y\|}$. Also define $\rho := \langle \hat{e}, \hat{y} \rangle$. Note that by Cauchy–Schwarz, we have $|\rho| \leq 1$.⁵

Special case: $|\rho| = 1$. Define $\Phi x = \rho \frac{\|y\|}{\|e\|} x$ and obtain:

$$\left\langle \begin{bmatrix} x \\ \Phi x \end{bmatrix}, M \begin{bmatrix} x \\ \Phi x \end{bmatrix} \right\rangle = \frac{\|x\|^2}{\|e\|^2} \left\langle \begin{bmatrix} e \\ y \end{bmatrix}, M \begin{bmatrix} e \\ y \end{bmatrix} \right\rangle \geq 0.$$

⁵Recall that in general, inner products are elements of \mathbb{F} , so ρ may be a complex number.

General case: $|\rho| < 1$. Due to (11), we have $M \neq 0$. So there must exist some $\eta \in \mathbb{F}$ such that $\begin{bmatrix} 1 \\ \eta \end{bmatrix}^* M \begin{bmatrix} 1 \\ \eta \end{bmatrix} \geq 0$. For any $x \in \mathcal{V}$, apply the projection theorem to decompose $x = x_{ey} + x_\perp$, where x_{ey} is a linear combination of \hat{e} and \hat{y} and x_\perp is orthogonal to both \hat{e} and \hat{y} . This yields

$$\begin{aligned} x_{ey} &:= \left(\frac{\langle \hat{e}, x \rangle - \rho \langle \hat{y}, x \rangle}{1 - |\rho|^2} \right) \hat{e} + \left(\frac{\langle \hat{y}, x \rangle - \bar{\rho} \langle \hat{e}, x \rangle}{1 - |\rho|^2} \right) \hat{y}, \\ x_\perp &:= x - x_{ey}. \end{aligned}$$

Note that if $x = e$, we have $e_{ey} = e$ and $e_\perp = 0$. We also have $\|x\|^2 = \|x_{ey}\|^2 + \|x_\perp\|^2$. Define the unit vectors

$$\hat{e}_\perp := \frac{\hat{y} - \rho \hat{e}}{\sqrt{1 - |\rho|^2}} \quad \text{and} \quad \hat{y}_\perp := \frac{\bar{\rho} \hat{y} - \hat{e}}{\sqrt{1 - |\rho|^2}}.$$

The vectors \hat{e}_\perp and \hat{y}_\perp are orthogonal to \hat{e} and \hat{y} , respectively. Write $M_{12} = |M_{12}|e^{i\varphi}$ (polar decomposition). Since $M_{21} = \overline{M_{12}}$, we have: $e^{-2i\varphi}M_{12} = M_{21}$. Finally, define Φ as

$$\Phi x := \frac{\|y\|}{\|e\|} \left(\langle \hat{e}, x_{ey} \rangle \hat{y} + e^{-2i\varphi} \langle \hat{e}_\perp, x_{ey} \rangle \hat{y}_\perp \right) + \eta x_\perp.$$

The function Φ is linear and using the fact that $e_{ey} = e$ and $e_\perp = 0$, it follows that $\Phi e = y$. Moreover, one can check that $\|\Phi x_{ey}\| = \frac{\|y\|}{\|e\|} \|x_{ey}\|$ and $\text{Re}(M_{12} \langle x_{ey}, \Phi x_{ey} \rangle) = \frac{\|y\|}{\|e\|} \|x_{ey}\|^2 \text{Re}(M_{12}\rho)$.

Thus, $\begin{bmatrix} x \\ \Phi x \end{bmatrix} = \begin{bmatrix} x_{ey} \\ \Phi x_{ey} \end{bmatrix} + \begin{bmatrix} x_\perp \\ \eta x_\perp \end{bmatrix}$ and $\left\langle \begin{bmatrix} x \\ \Phi x \end{bmatrix}, M \begin{bmatrix} x \\ \Phi x \end{bmatrix} \right\rangle$

$$= \left\langle \begin{bmatrix} x_{ey} \\ \Phi x_{ey} \end{bmatrix}, M \begin{bmatrix} x_{ey} \\ \Phi x_{ey} \end{bmatrix} \right\rangle + \left\langle \begin{bmatrix} x_\perp \\ \eta x_\perp \end{bmatrix}, M \begin{bmatrix} x_\perp \\ \eta x_\perp \end{bmatrix} \right\rangle.$$

The first term simplifies to: $\left\langle \begin{bmatrix} x_{ey} \\ \Phi x_{ey} \end{bmatrix}, M \begin{bmatrix} x_{ey} \\ \Phi x_{ey} \end{bmatrix} \right\rangle$

$$\begin{aligned} &= M_{11} \|x_{ey}\|^2 + 2 \text{Re}(M_{12} \langle x_{ey}, \Phi x_{ey} \rangle) + M_{22} \|\Phi x_{ey}\|^2 \\ &= \|x_{ey}\|^2 \left(M_{11} + 2 \text{Re}(M_{12}\rho) \frac{\|y\|}{\|e\|} + M_{22} \frac{\|y\|^2}{\|e\|^2} \right) \\ &= \frac{\|x_{ey}\|^2}{\|e\|^2} (M_{11} \|e\|^2 + 2 \text{Re}(M_{12} \langle e, y \rangle) + M_{22} \|y\|^2) \\ &= \frac{\|x_{ey}\|^2}{\|e\|^2} \left\langle \begin{bmatrix} e \\ y \end{bmatrix}, M \begin{bmatrix} e \\ y \end{bmatrix} \right\rangle \geq 0. \end{aligned}$$

The second term simplifies to

$$\left\langle \begin{bmatrix} x_\perp \\ \eta x_\perp \end{bmatrix}, M \begin{bmatrix} x_\perp \\ \eta x_\perp \end{bmatrix} \right\rangle = \|x_\perp\|^2 \begin{bmatrix} 1 \\ \eta \end{bmatrix}^* M \begin{bmatrix} 1 \\ \eta \end{bmatrix} \geq 0.$$

Therefore, we have $\left\langle \begin{bmatrix} x \\ \Phi x \end{bmatrix}, M \begin{bmatrix} x \\ \Phi x \end{bmatrix} \right\rangle \geq 0$, as required. \blacksquare

References

- [1] B. D. O. Anderson. The small-gain theorem, the passivity theorem and their equivalence. *Journal of the Franklin Inst.*, 293(2):105–115, 1972.
- [2] L. J. Bridgeman and J. R. Forbes. The extended conic sector theorem. *IEEE Trans. Autom. Control*, 61(7):1931–1937, 2016.
- [3] L. J. Bridgeman and J. R. Forbes. A comparative study of input-output stability results. *IEEE Trans. Autom. Control*, 63(2):463–476, 2018.
- [4] R. Brockett. The status of stability theory for deterministic systems. *IEEE Trans. Autom. Control*, 11(3):596–606, 1966.
- [5] J. Carrasco and P. Seiler. Conditions for the equivalence between IQC and graph separation stability results. *Int. J. Control*, 92(12):2899–2906, 2019.
- [6] J. B. Conway. *A course in functional analysis*. Springer-Verlag, New York, 1990.
- [7] S. Cyrus. *Stability of Interconnected Sector-bounded Systems, with Application to Designing Optimization Algorithms*. PhD thesis, University of Wisconsin–Madison, 2021.
- [8] S. Cyrus and L. Lessard. Unified necessary and sufficient conditions for the robust stability of interconnected sector-bounded systems. In *IEEE Conf. Decision Contr.*, pages 7690–7695, 2019.
- [9] C. A. Desoer and M. Vidyasagar. *Feedback systems: input-output properties*, volume 55. SIAM, 1975.
- [10] M. Fu, S. Dasgupta, and Y. C. Soh. Integral quadratic constraint approach vs. multiplier approach. *Automatica*, 41(2):281–287, 2005.
- [11] M. R. Hestenes. *Optimization theory: The finite dimensional case*. Wiley, 1975.
- [12] S. Z. Khong and C.-Y. Kao. Converse theorems for integral quadratic constraints. *IEEE Trans. Autom. Control*, 66(8):3695–3701, 2020.
- [13] S. Z. Khong and A. J. van der Schaft. On the converse of the passivity and small-gain theorems for input-output maps. *Automatica*, 97:58–63, 2018.
- [14] A. Lur’e and V. Postnikov. On the theory of stability of control systems. *Applied mathematics and mechanics*, 8(3):246–248, 1944.
- [15] M. J. McCourt and P. J. Antsaklis. Control design for switched systems using passivity indices. In *Amer. Control Conf.*, pages 2499–2504, 2010.
- [16] A. Megretski. Necessary and sufficient conditions of stability: A multiloop generalization of the circle criterion. *IEEE Trans. Autom. Control*, 38(5):753–756, 1993.
- [17] A. Megretski and A. Rantzer. System analysis via integral quadratic constraints. *IEEE Trans. Autom. Control*, 42(6):819–830, 1997.
- [18] A. Megretski and S. Treil. Power distribution inequalities in optimization and robustness of uncertain systems. *J. Math. Syst., Estimation Contr.*, 3(3):301–319, 1993.
- [19] H. Pfifer and P. Seiler. Integral quadratic constraints for delayed nonlinear and parameter-varying systems. *Automatica*, 56:36–43, 2015.
- [20] M. G. Safonov. *Stability and robustness of multivariable feedback systems*. MIT press, 1980.
- [21] P. Seiler. Stability analysis with dissipation inequalities and integral quadratic constraints. *IEEE Trans. Autom. Control*, 60(6):1704–1709, 2014.
- [22] J. S. Shamma. The necessity of the small-gain theorem for time-varying and nonlinear systems. *IEEE Trans. Autom. Control*, 36(10):1138–1147, 1991.
- [23] A. R. Teel. On graphs, conic relations, and input-output stability of nonlinear feedback systems. *IEEE Trans. Autom. Control*, 41(5):702–709, 1996.
- [24] A. R. Teel, T. T. Georgiou, L. Praly, and E. D. Sontag. Input-output stability. In *The control handbook, second edition: Control system advanced methods*, chapter 44. CRC Press, 2011.

- [25] A. J. van der Schaft. *L₂-Gain and passivity techniques in nonlinear control, Third edition*. Springer, 2017.
- [26] J. Veenman, C. W. Scherer, and H. Köroğlu. Robust stability and performance analysis based on integral quadratic constraints. *European Journal of Control*, 31:1–32, 2016.
- [27] M. Vidyasagar. *Nonlinear Systems Analysis*. Society for Industrial and Applied Mathematics, 2nd edition, 2002.
- [28] J. C. Willems. Dissipative dynamical systems—Part II: Linear systems with quadratic supply rates. *Archive for rational mechanics and analysis*, 45(5):352–393, 1972.
- [29] V. Yakubovich. S-procedure in nonlinear control theory. *Vestnik Leningrad Univ. Math.*, 4:73–93, 1997.
- [30] G. Zames. On the input-output stability of time-varying nonlinear feedback systems—Part I: Conditions derived using concepts of loop gain, conicity, and positivity. *IEEE Trans. Autom. Control*, 11(2):228–238, 1966.
- [31] G. Zames. On the input-output stability of time-varying nonlinear feedback systems—Part II: Conditions involving circles in the frequency plane and sector nonlinearities. *IEEE Trans. Autom. Control*, 11(3):465–476, 1966.
- [32] K. Zhou, J. C. Doyle, and K. Glover. *Robust and optimal control*. Prentice Hall, Upper Saddle River, N.J, 1996.